

Accurate Field Shape Model for Beam-Wave Interaction Simulation of Folded-Waveguide Traveling-Wave Tubes

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Session 2.3 - Traveling Wave Tubes - September 9, 2016

TUHH

THALES

- ① Motivation
- ② Theory
- ③ Application
- ④ Conclusion

① Motivation

② Theory

③ Application

④ Conclusion



- Increasing number of services requiring high data rates
 - Use higher bands as carrier frequencies
 - Increased free-space and atmospheric loss

Images from “energyworx.com” and “a-read.com”.



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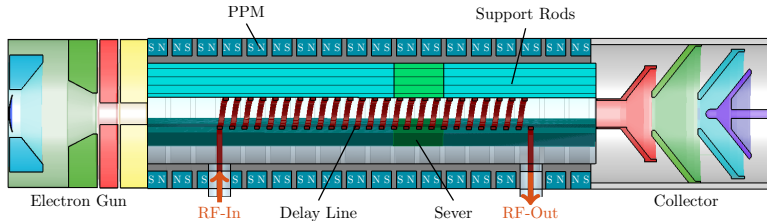
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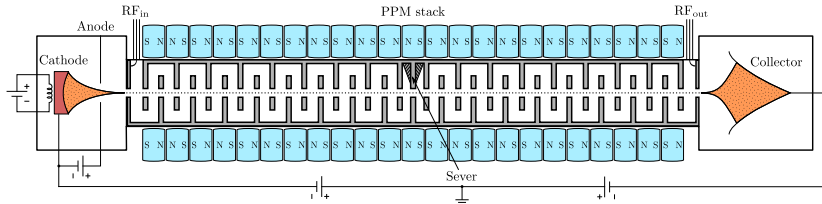
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The Traveling-Wave Tube Amplifier



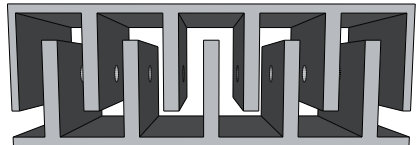
- Conversion of kinetic to RF energy
- State-of-the-art technology for satellite communication
- Thermal/mechanical limitations of commonly used helix delay lines above 40 GHz

The Folded-Waveguide Delay Line



Pros and Cons

- + Ease of fabrication
- + Large bandwidth
- + Good thermal properties
- Weak beam-wave coupling



Simulate Before You Fabricate

Large time/monetary costs of fabrication

- Avoid trial-and-error productions
- Characterize device at the PC level
- Optimize performance beforehand

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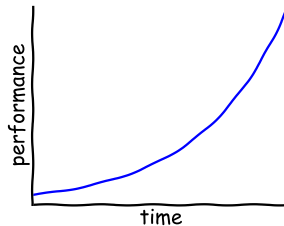
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Generalized vs. Specialized Software

Generalized Tool

- + **Predict performance of any new/exotic device**
- + Exploration of new concepts
- + Analysis of phenomena
- Limited suitability for design

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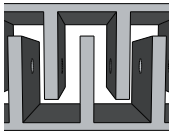


① Motivation

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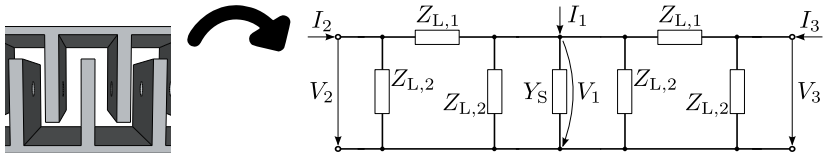
④ Conclusion



- Electromagnetic waves propagate along delay line
- Dispersive properties are mapped to lumped-element circuit¹
- Voltage and current amplitudes define waves

¹T. M. Antonsen, A. N. Vlasov, D. P. Chernin, I. A. Chernyavskiy, and B. Levush: "Transmission Line Model for Folded Waveguide Circuits", *IEEE Trans. Electron Devices*, vol. 60, pp. 2906-2911, 2013

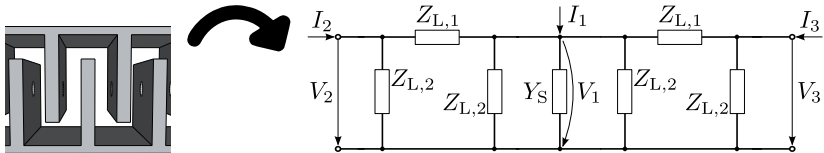
Equivalent Circuit Description



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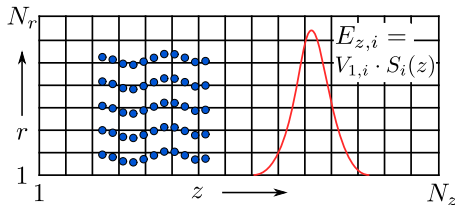
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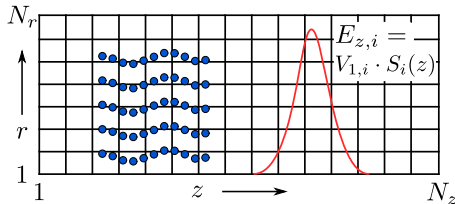


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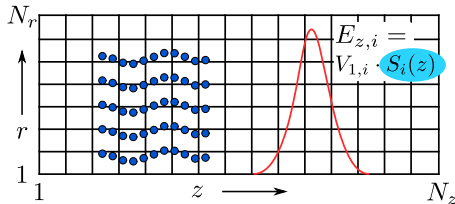


- Discrete macro-particles propagate inside a discretized domain
- Solution of Lorenz force equation $\ddot{\mathbf{x}} = -q/m_q \cdot \mathbf{E}$
- Field shape function defines link between fields and amplitudes



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Particle-in-Cell Approach



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Significance of Axial Electric Field

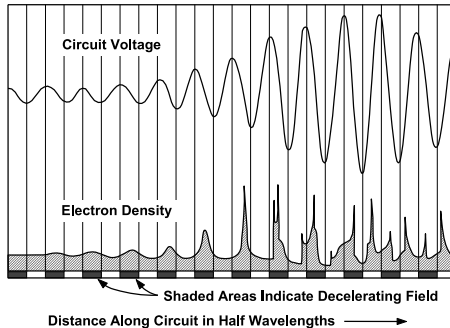


Figure: Illustration of field strength and charge density².

- Electric field induces beam modulation and counteracts de-bunching forces

→ Influence on maximum achievable efficiency

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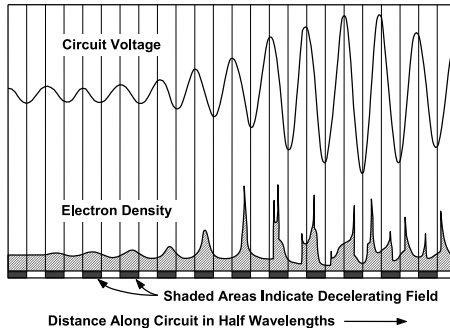


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About Spatial Harmonics

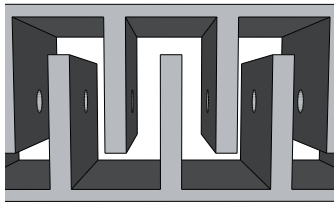


Figure: FW delay line.

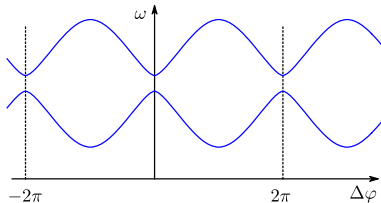


Figure: Dispersion diagram.

- Periodic structure supports propagating modes
- EM fields are superposition of spatial harmonics
- All harmonics have same group but different phase velocity
- Amplitude of synchronous harmonic defines beam-wave coupling

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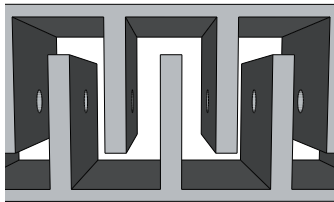


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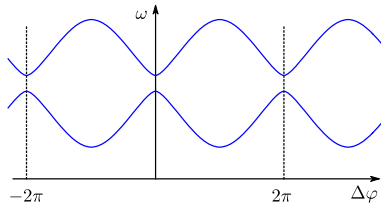


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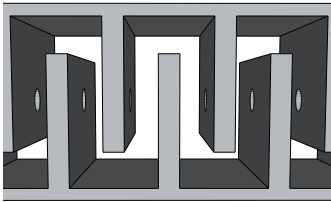


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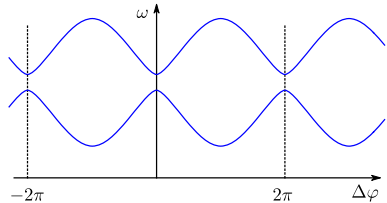


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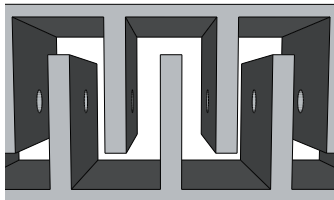


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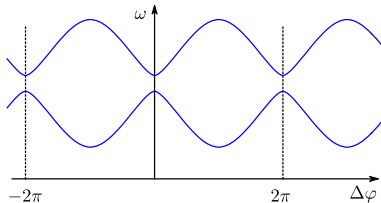


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- Coupling impedance $R_{c,n}$ defined by Pierce³
 - Strength of n th space-harmonic amplitude independent of transported power
- Kino impedance R_{Kino} defined by Curnow⁴
 - Measure of total field strength (all spatial harmonics) independent of transported power

→ Related by shape of axial electric field

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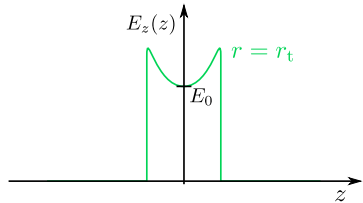
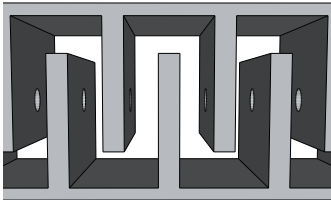
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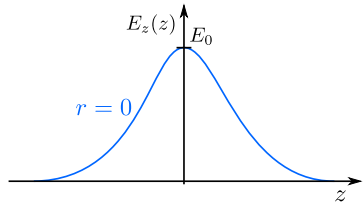
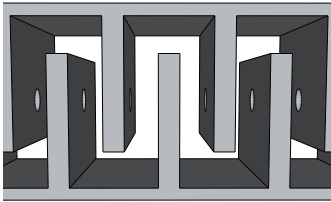
Hyperbolic Field Shape Model



- At beam tunnel radius axial field is zero on the metal and has hyperbolic features in the gap⁵
- Unknown curve parameter of hyperbolic shape
- Sudden change of field distribution at gap corner

⁵H. G. Kosmahl and G. M. Branch, Jr., "Generalized Representation of Electric Fields in Interaction Gaps of Klystrons and Traveling-Wave Tubes", *IEEE Trans. Electron Devices*, vol. 20, no. 7, pp. 621-629, 1973.

Field Shape From Laplace Equation



- Axial field at beam center has features similar to a Gaussian
- Evanescent fields in beam tunnel (waveguide below cut-off)
- Extract from solution of Laplace equation

Comparison of Shape Models

Hyperbolic cosine

- + Analytical expression
- Fixed far from beam axis
- More information necessary

Laplace equation

- + Depends only on geometry
- + Computationally inexpensive
- + Shape outside axis included

→ Harmonic amplitudes are not controlled in either model!

Hyperbolic cosine

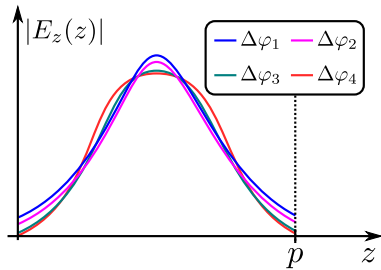
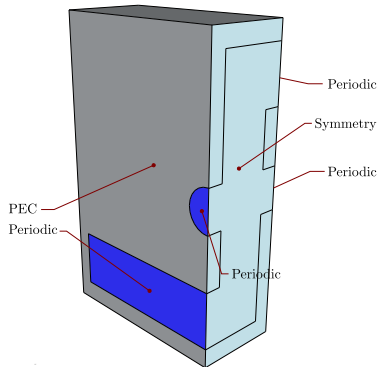
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Synthesis of Shape Function



- Periodic unit cell with corresponding boundary conditions
- Eigenfrequencies and -modes for each phase advance $\Delta\varphi$
- Extract complex axial electric field $E_z(z, \beta)$

Periodic field in unit cell

$$E_z(z, \beta) = V_\beta \cdot \sum_{m=-\infty}^{\infty} S(z - mp) \cdot e^{-jm\beta p}$$

... some time and ink later ...

⁶Similar approach to [N. M. Ryskin and V. N. Titov](#), "Nonstationary Simulation of Electron Beam Interaction with Coupled Resonant Microwave Oscillators", *Proc. IEEE Int. Conf. Plasma Science (ICOPS)*, 2007, 725.

Derivation of the Shape⁶

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Local field shape

$$S(z) = \frac{p}{2\pi} \int_0^{2\pi/p} \frac{E_z(z, \beta)}{V_\beta} d\beta$$

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Superposition of Space Harmonics

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Synthesis formula

$$S(z) = \frac{1}{\pi} \int_0^\infty \beta p \sqrt{\frac{R_{c,\beta}}{R_{\text{Kino},\beta}}} \cdot \cos(\beta z) d\beta$$

- + Shape is related to geometry
- + Correct representation of harmonic amplitudes
- Integration needs to be truncated
- Eigenmode simulation required

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Radial Extrapolation of Field Shape

Radial wavenumber

$$\gamma^2 = \beta^2 - k^2$$

- Local cylindrical symmetry

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Axial/Radial shape functions

$$S_z(z, r) = \frac{1}{\pi} \int_0^\infty I_0(\gamma r) \cdot \beta p \sqrt{\frac{R_{c,\beta}}{R_{\text{Kino},\beta}}} \cdot \cos(\beta z) d\beta$$

$$S_r(z, r) = \frac{1}{\pi} \int_0^\infty \left(\frac{\beta}{\gamma}\right) \cdot I_1(\gamma r) \cdot \beta p \sqrt{\frac{R_{c,\beta}}{R_{\text{Kino},\beta}}} \cdot \cos(\beta z) d\beta$$

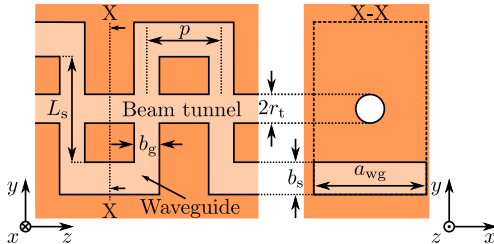
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Example Geometry

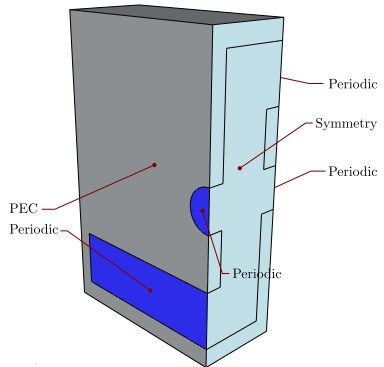
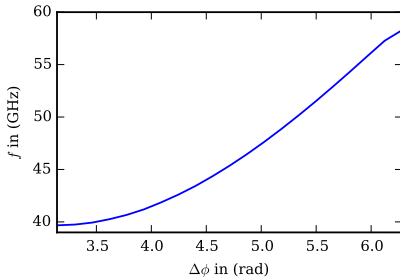


Geometry parameters in (mm).

p	r_t	L_s	a_{wg}	b_s	b_g
1.1	0.35	2.0	3.8	0.55	0.55

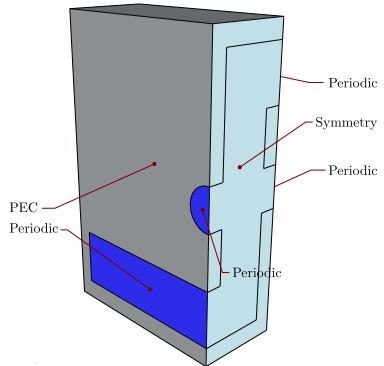
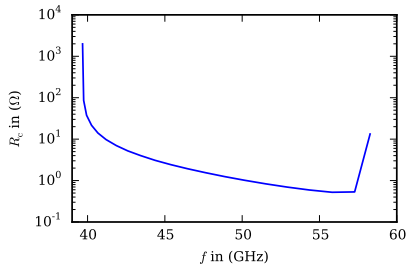
- Conventional FW delay line
- Homogeneous cross section
- Relatively small beam tunnel

Eigenmode Simulation Results

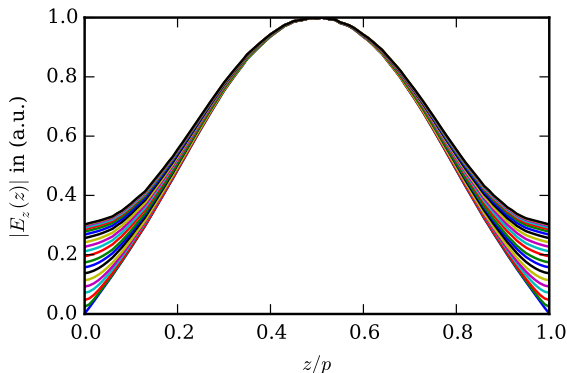


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- Passband from approx. 40 GHz to 58 GHz
- Coupling impedance between $1\ \Omega$ and $10\ \Omega$

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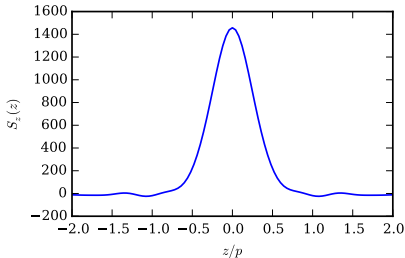


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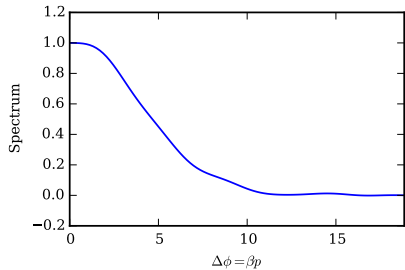
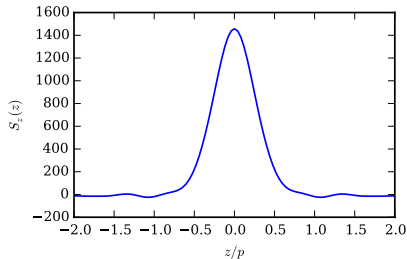
- Axial field varies with frequency, or rather, phase advance
- Actual shape is significantly different

Extracted Field Shape and Spectrum



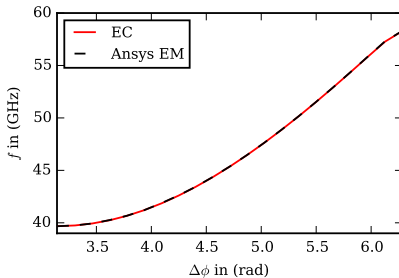
- Strongly localized field shape
- Small contribution to adjacent cells ($z = \pm p$)
- Spectrum largest for $\Delta\phi \leq 2\pi$

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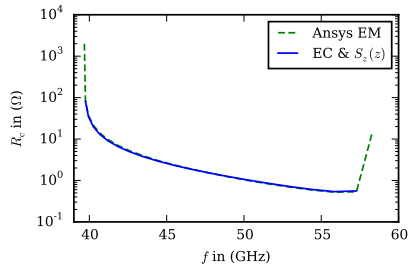
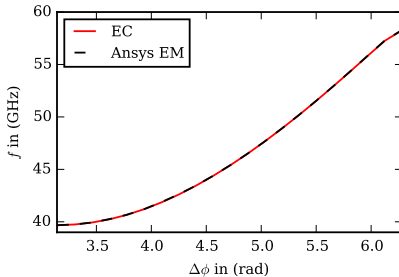
Comparison of Results



- Equivalent circuit fits dispersion nicely
- With $S_z(z)$ the coupling impedance is well represented
- Error below 10 % across band suitable for beam-wave interaction computation⁷

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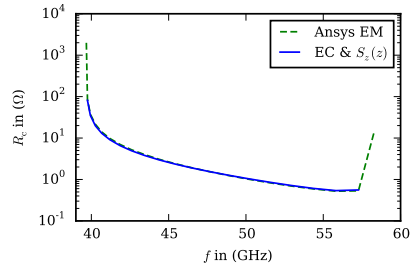
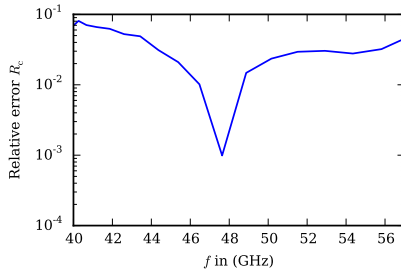
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- Correct amplitudes of space harmonics
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Outlook

- Implementation in beam-wave interaction algorithm
- Applicability to other delay line topologies

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Thank you for your attention!

Acknowledgments

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